



# MAX77812 PCB Layout Guide

UG6819 Rev 0; 2/19

## Abstract

*The MAX77812 is a quad-phase, high-current buck converter for high-end gaming consoles, VR/AR headsets, DSLR cameras, drones, network switches and routers, and FPGA systems that use multicore processors. The multiphase buck architecture reduces the size of the external components while providing the ripple current cancellation and world-class transient response. Differential remote sensing feedback enables tight DC and AC accuracy at the point of load. The MAX77812's flexible architecture allows five user-selectable phase configurations: 4, (3+1), (2+2), (2+1+1), and (1+1+1+1).*

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## General Guidelines

- The power components should be placed first, and then small analog control signals.
- It is important to always have a ground layer next to the power stage layer. A solid ground layer provides an uninterrupted ground return path between the input and the output capacitors during switch ON time. A solid plane minimizes inductance to the absolute minimum, is a very good thermal conductor, and can act as a heat sink.
- It is recommended to have thick copper for the external high-current power layers to minimize PCB conduction loss and thermal impedance.
- The power stage loop (**Figure 1**) that is made by the LX traces, inductor (L), and output capacitor ( $C_{OUT}$ ) coming back to the PGNDx bumps should be minimized for EMC considerations.

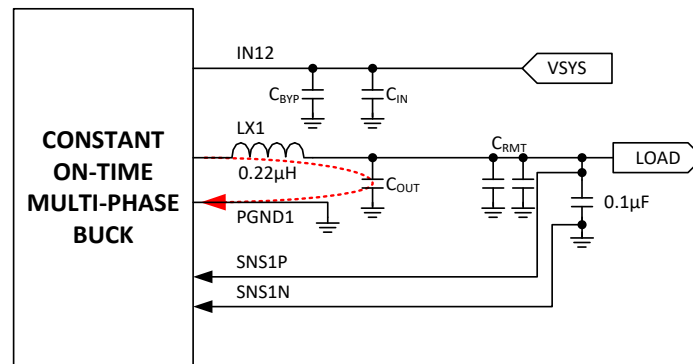


Figure 1. Buck power stage loop.

- The input capacitors ( $C_{IN}$  and  $C_{BYP}$ ) should be located close to the input bumps of each phase.
- Analog ground (AGND), digital ground (DGND), and power ground (PGND) bumps should be directly connected to the ground plane separately in order to avoid a common impedance ground.
- It is important to have impedance matching between phases for stable operation in a multiphase configuration.  
**Note:** The output PCB trace of each phase should be as symmetric as possible.
- The negative output voltage sensing bumps ( $SNSx\_N$ ) should NOT be connected to the power ground bumps (PGNDx) on the top layer.  
**Note:** The power ground bumps should be tied on the ground plane.
- For multiphase configurations, the output voltage sensing bumps (differential pair) for the master phase should be connected to the middle point of the output phases.
- The output voltage sensing bumps (differential pair) for the slave phases must still be connected to their outputs for determining the on-time of switching cycles.  
**Note:** The output voltage sensing signals of slave phases can be combined with the master phase at the bumps.
- The output voltage sensing trace should not intersect the power stage loop.
- For improving load-transient response, additional capacitors ( $C_{RMT}$ ) can be added between the output capacitor ( $C_{OUT}$ ) and the output voltage sensing point (optional).

## Power Stage

Low-ESR ceramic bypass capacitors should be located as close to the input bumps as possible. Generally,  $0.1\mu\text{F}$  to  $10\mu\text{F}$  capacitors are suitable for effective decoupling of high-frequency power-supply noise. Crosstalk is essentially a capacitive coupling event that occurs when a noise signal is induced to nearby traces through parasitic low-impedance paths. It is recommended to keep the distance between adjacent PH1 and PH2 traces (for dual-phase configuration) at least two to three times their width apart. The power stage loop should be minimized for EMC considerations. It is important to have thick copper traces for the outer high-current power layers to minimize PCB conduction loss and improve the heat dissipation. Multiple vias are recommended for all paths that carry high currents (PGND,  $V_{\text{IN}}$ , and LX). Placement of vias should create the shortest possible current loops and must not obstruct the flow of currents or mirror currents in the ground plane.

For the multiphase configurations, the impedance matching between phases is important for stable operation. The output trace of each phase should be as symmetric as possible. The output voltage sensing bumps (differential pair) for the master phase should be connected to the middle point of the output phases, as shown in **Figure 2**. The output voltage sensing bumps for slave phase(s) must be connected to their output for determining the on-time. The output voltage sensing signals of slave phase(s) can be tied with the corresponding master phase(s), as shown in **Figure 3**.

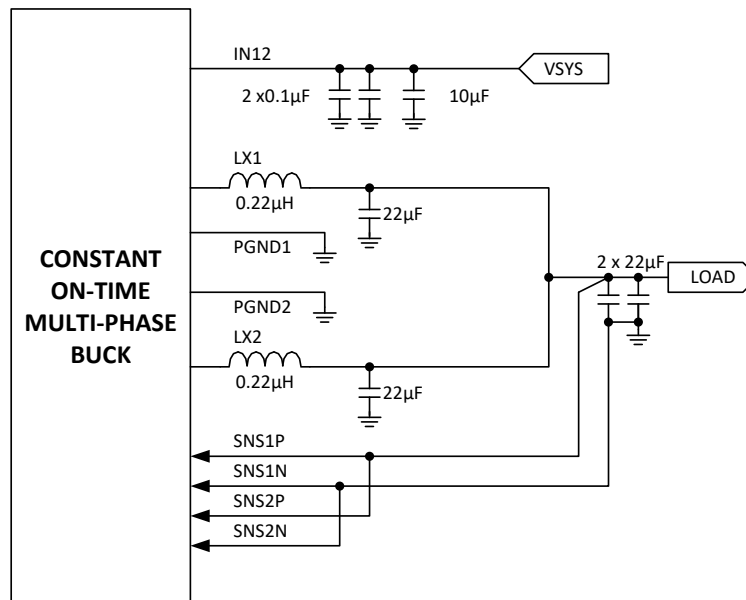


Figure 2. Dual-phase buck schematic and output voltage sensing point.

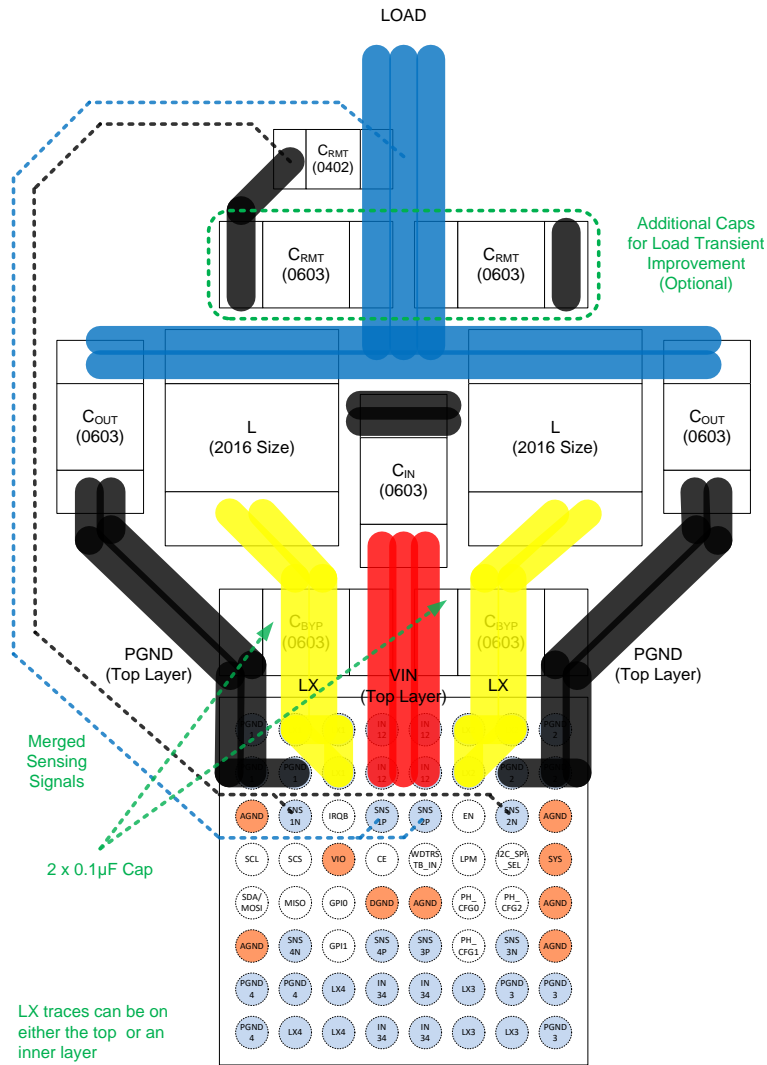
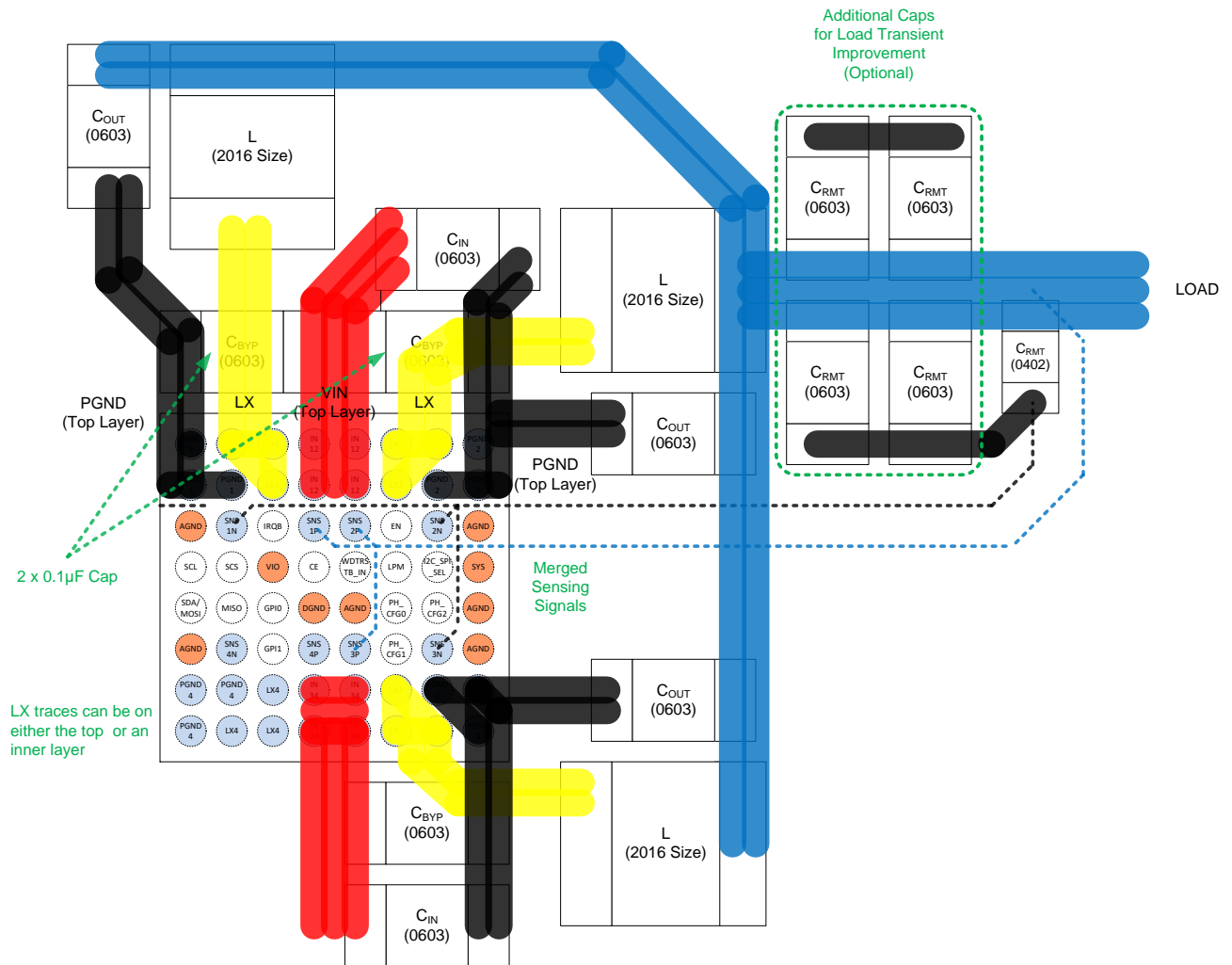


Figure 3. Dual-phase PCB layout recommendation.

**Table 1** shows recommended capacitors for achieving optimal noise performance in dual-phase PCB layout. See **Figure 4** and **Figure 5** for the recommended PCB layout for 3-phase and quad-phase configurations.

**Table 1. Recommended Capacitors for Dual-Phase Configuration**

	Nominal Capacitance	Remark
Bypass Capacitor	0.1 $\mu$ F (50V X5R, 0603)	Reduces crosstalk (switching spikes) between PH1 and PH2; must place input bypass capacitors as close as possible to input bumps
Input Capacitor	10 $\mu$ F (6.3V X5R, 0603)	Needed for input voltage ripple reduction
Local Output Capacitor	22 $\mu$ F (6.3V X5R, 0603)	Needed for stable operation
Remote Capacitor	Up to 10 $\mu$ F per phase	Optional for better output ripple and load-transient response



*Figure 4. Recommendation for 3-phase PCB layout.*

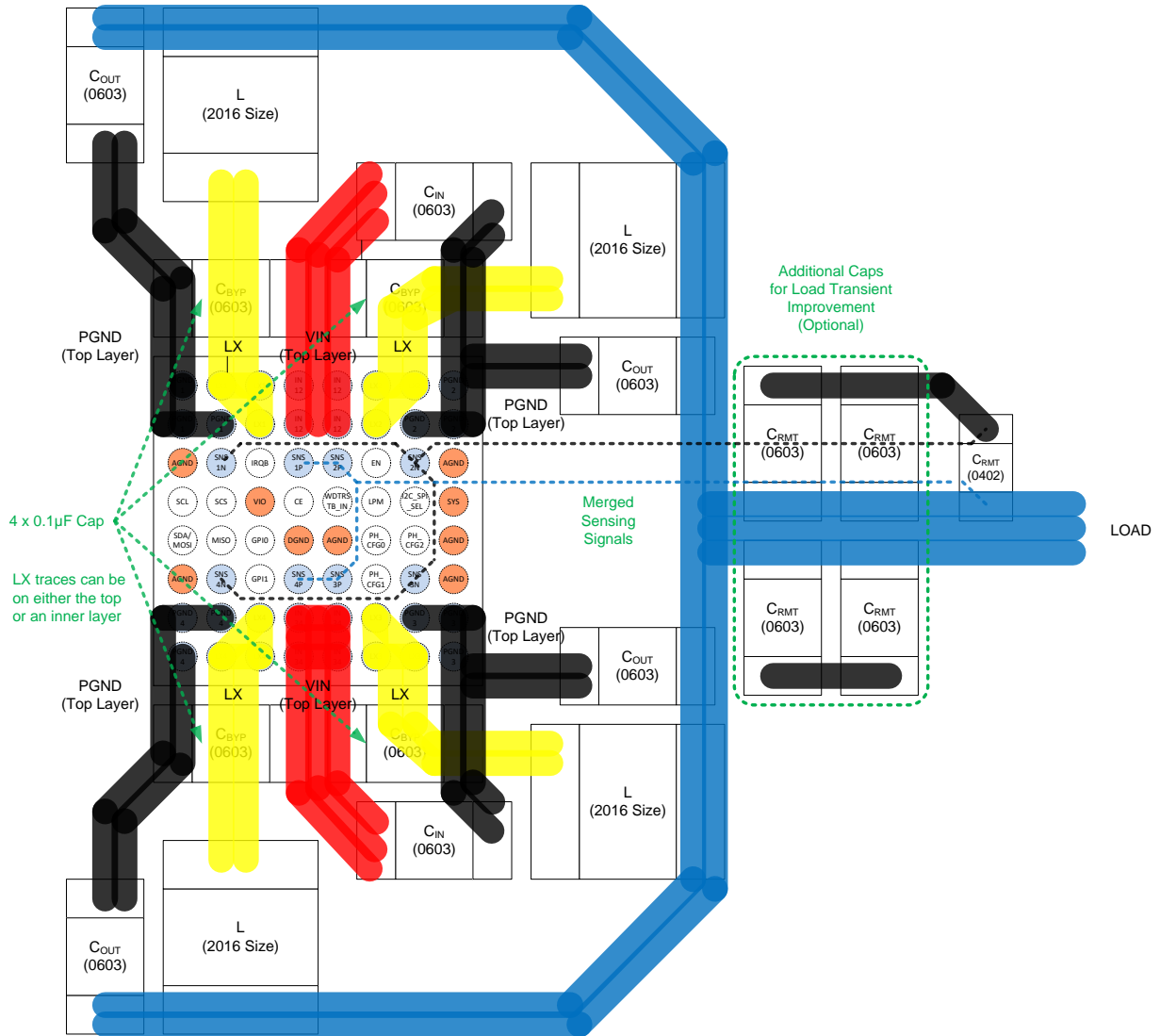


Figure 5. Quad-phase PCB layout recommendation.

## PCB Layout Examples and Simulation Results

**Figure 6** shows examples of a dual-phase PCB layout. With balanced output voltage sensing, the switching waveforms are regular, resulting in smaller output voltage ripple as shown in **Figure 7**. **Figure 8** shows that unbalanced output voltage sensing gives higher output voltage ripple due to slight changes in off-time period cycle by cycle.

### PCB Layout Details—Common for Balanced and Unbalanced Cases

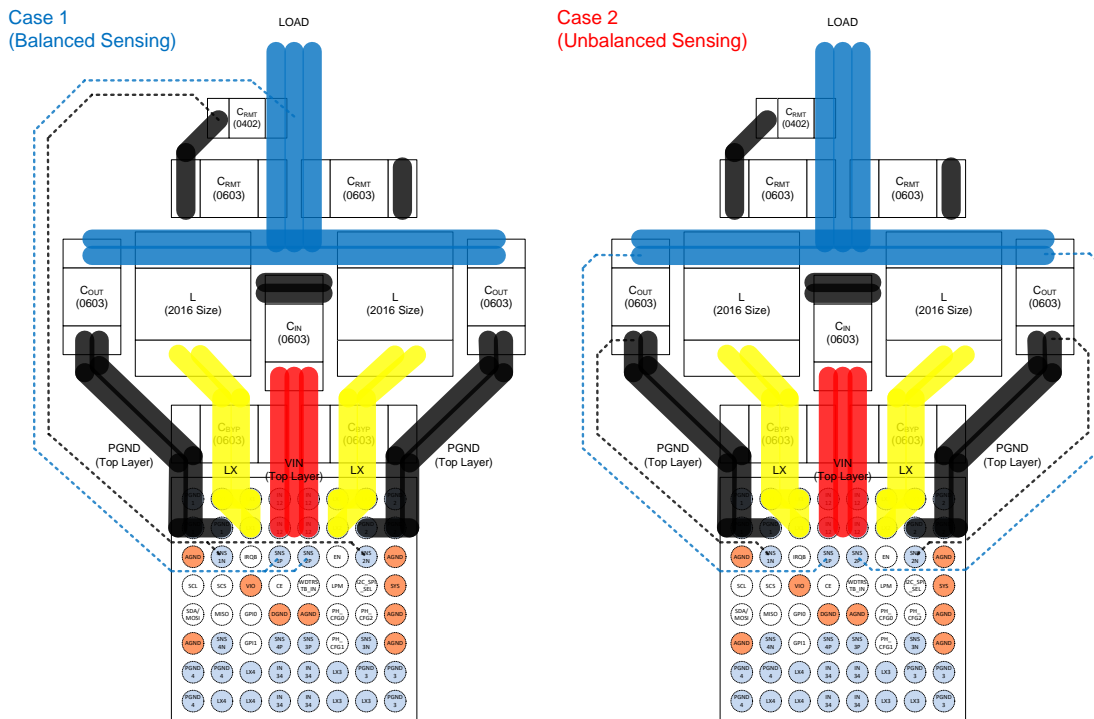
- PH1 and PH2 outputs are tied from  $C_{OUT}$  and connected to the load together.
- Parasitics of PH1 and PH2 traces are matched (symmetric PCB layout).
- Load is connected to the middle point of the output phases.

### Case 1—Balanced Output Sensing

The PH1 and PH2 output voltage sensing signals are tied together and connected to the middle point of the output phases ( $C_{RMT}$ ).

### Case 2—Unbalanced Output Sensing

The PH1 and PH2 output voltage sensing signals are connected to their own local output capacitors ( $C_{OUT}$ ) separately.



*Figure 6. Dual-phase PCB layouts with balanced (Case 1) and unbalanced (Case 2) sensing.*



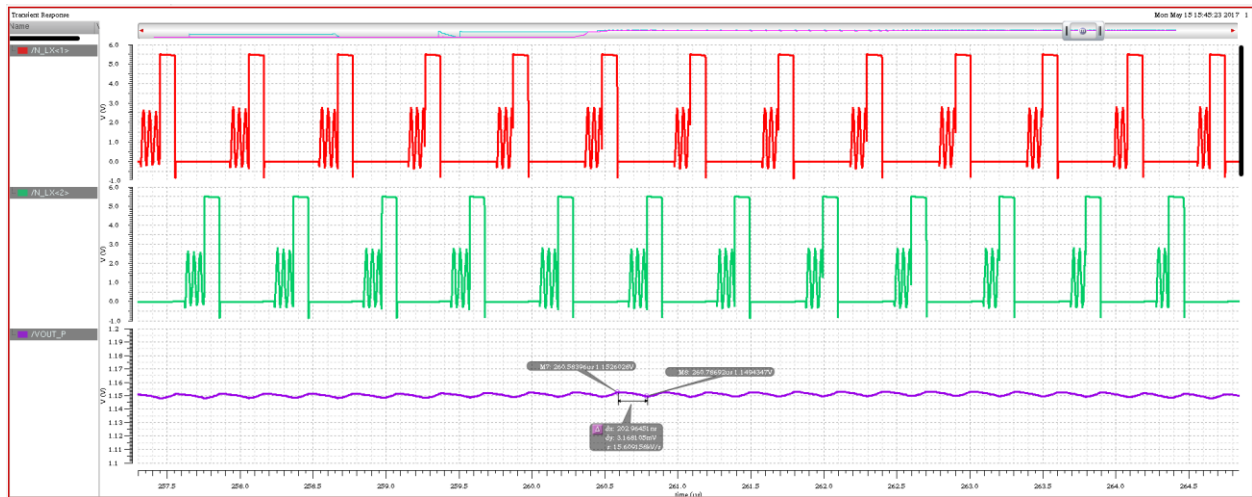


Figure 7. Switching waveforms with balanced output sensing example (Case 1).

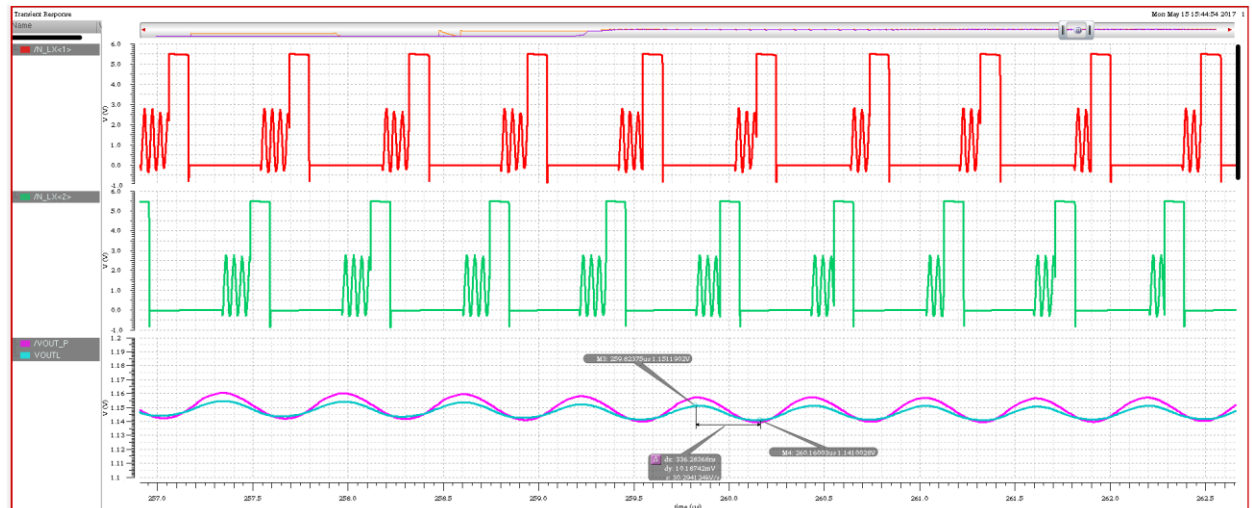


Figure 8. Switching waveforms with unbalanced output sensing example (Case 2).

# PCB Layout Example

LAMINATION DIAGRAM				
LAYER NUMBER	LAYER NAME	FINISHED CU WEIGHT (oz)	DIELECTRIC THICKNESS (in.)	DIELECTRIC MATERIAL
1	TOP	2		FOIL
2	INTERNAL2	0.5	TBD	FR4(RoHS)/EQIV
3	INTERNAL3	0.5	TBD	FR4(RoHS)/EQIV
4	INTERNAL4	0.5	TBD	FR4(RoHS)/EQIV
5	INTERNAL5	0.5	TBD	FR4(RoHS)/EQIV
6	BOTTOM	2		FOIL

THE FINISHED PCB THICKNESS TO BE: 0.062 +/- 0.010

Figure 9. Lamination diagram.

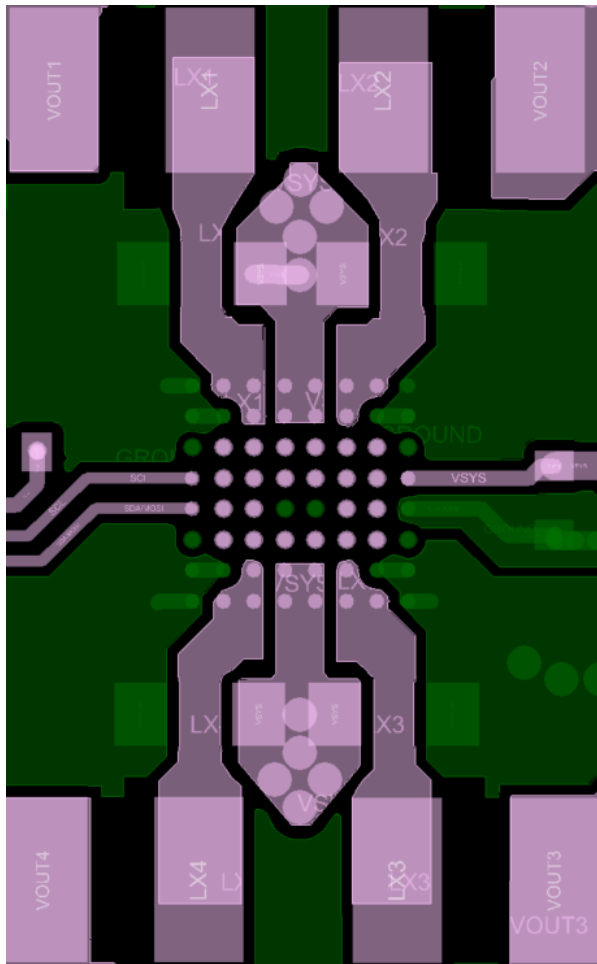


Figure 10. Top layer.

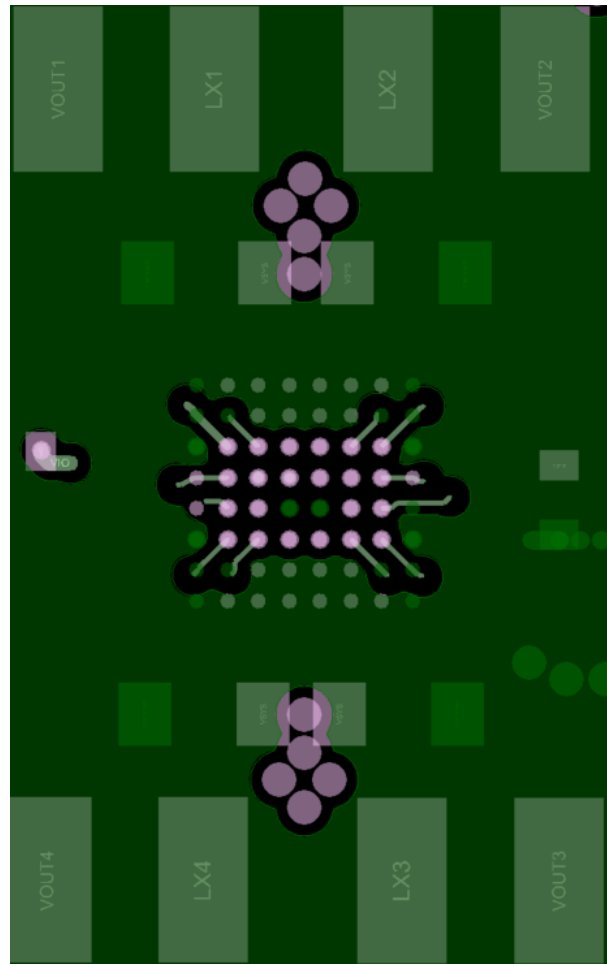


Figure 11. Second layer.

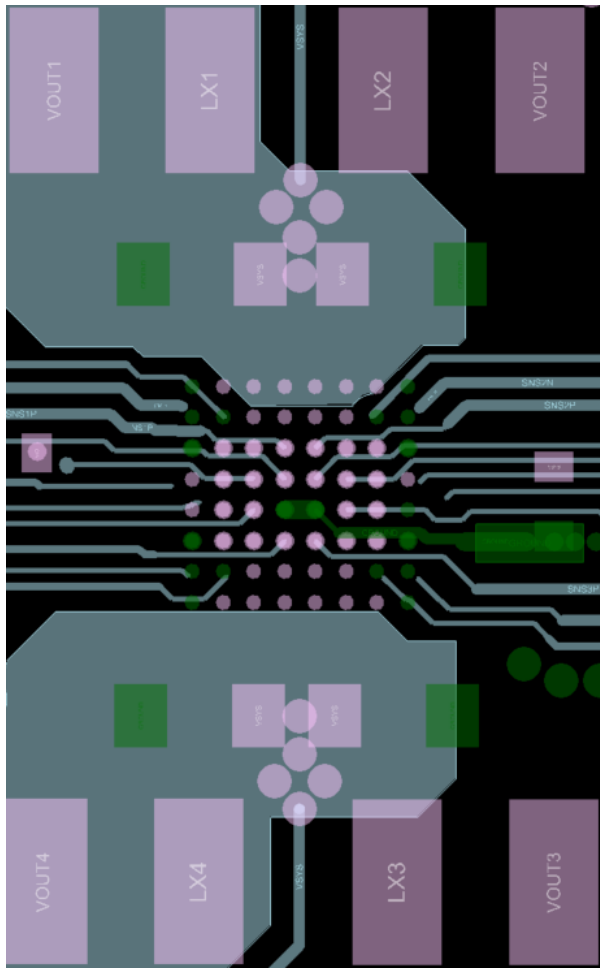


Figure 12. Third layer.

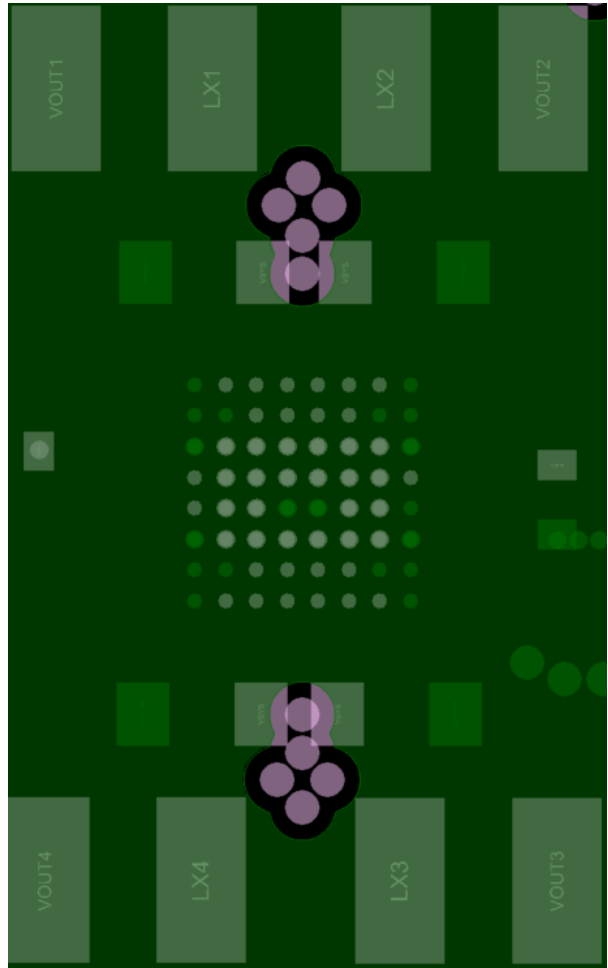


Figure 13. Fourth layer.

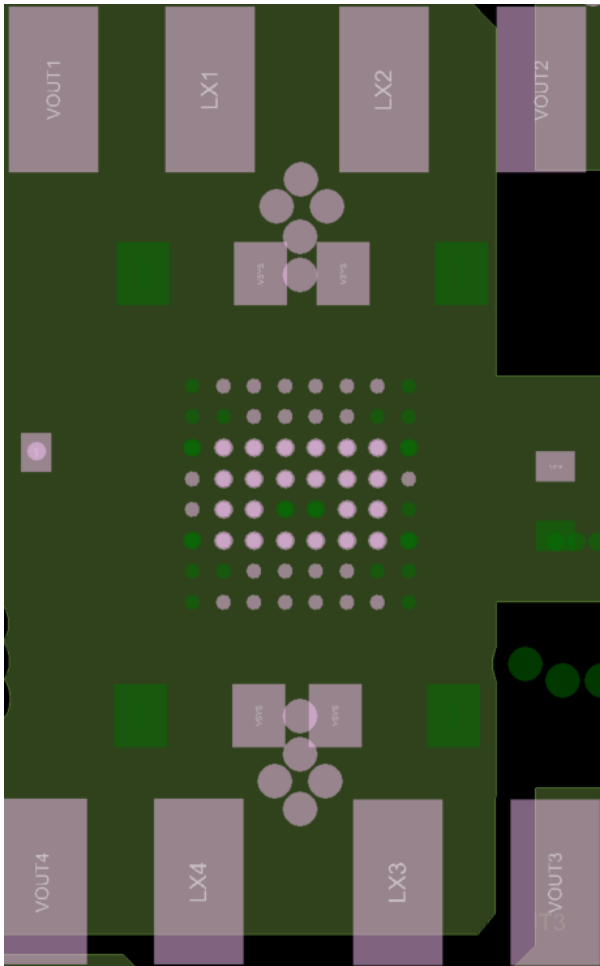


Figure 14. Fifth layer.

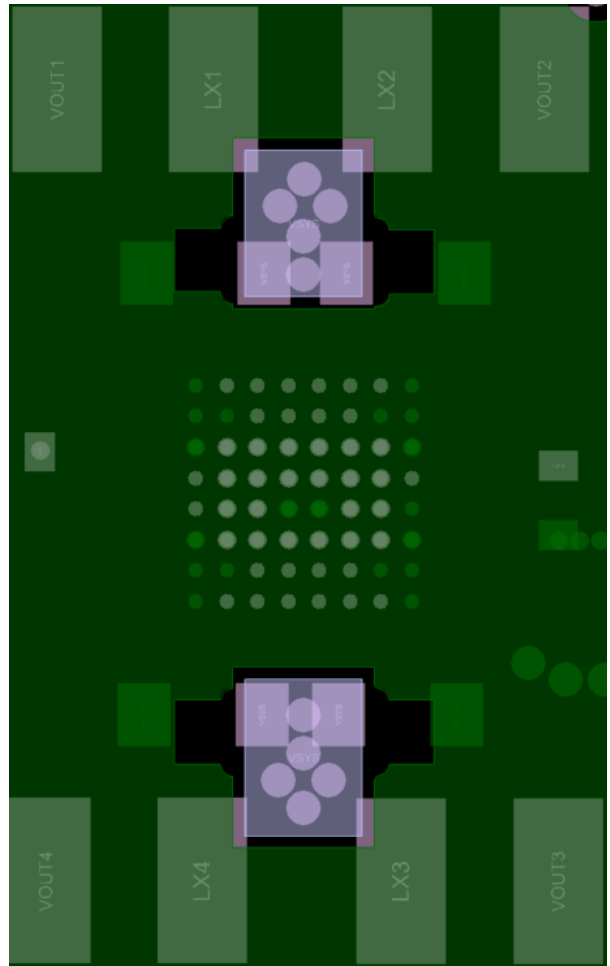


Figure 15. Bottom layer.

## Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	02/19	Initial release	—

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